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# Vector fitting approximation of a cylinder nonreflecting boundary kernel

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To employ the modal nonreflecting boundary condition (MNRBC) in cylindrical coordinates in the finite element time domain (FETD) method, a time domain kernel expression must be found that it is the inverse Laplace transform (ILT) of a known frequency domain function. The inverse Laplace transformation is achieved using a methodology based on the partial fraction expansion of the frequency domain function. However, to date, no FETD results have been published based on this MNRBC methodology. A simpler implementation of the methodology based on vector fitting (VF) is proposed. Using the VF approach, FETD-MNRBC results of plane wave scattering from a cylinder are presented for the first time.

*Introduction:* The modal nonreflecting boundary condition (MNRBC), in cylindrical coordinates, is a well established boundary condition in the two-dimensional (2D) finite element frequency domain (FEFD) method simulations [1]. This boundary condition is based on the fact that the scattered field on a circular (fictitious) boundary surrounding a cylinder of arbitrary cross-section can be expressed in terms of summation of modal functions, of integer modal orders  $n$ , which are products of Hankel functions or modified Bessel functions and azimuthal function terms, see for example [2]. The time domain version of this boundary condition has not been employed in finite element time domain (FETD) method simulations although the general methodology for developing a time domain MNRBC has been presented by Alpert et al [3]. The methodology relies on finding, for each order  $n$ , the time domain expression of a cylinder nonreflecting boundary kernel. This requires that for each order  $n$  the inverse Laplace transform (ILT) of a known function that appears in the MNRBC in the frequency domain is found. This is achieved by expressing the frequency domain function as a summation of partial fractions which via the ILT are expressed as a summation of exponential

terms in the time domain. Because of the complexity involved in implementing the partial fraction expansion in [3] and the fact that only a limited number of partial fraction coefficients that correspond to a few cylinder kernel modal orders ( $n=1,2,3,4$ ) were presented in [3], an alternative simpler approach of implementing the partial fraction expansion based on vector fitting (VF) is proposed using the publicly available software VECTFIT [4]. This VF approach may have a greater appeal among engineers. Through computations, it is demonstrated that the VF results are of comparable accuracy to those of Alpert et al [3]. In addition, FETD-MNRBC results, based on VF, are presented for the first time.

*The VF approach:* To demonstrate the proposed VF partial fraction expansion approach and its accuracy, the Laplace transform expression  $Q_n(s)$  of the time domain cylinder nonreflecting boundary kernel  $q_n(t)$ , used by Alpert et al (eq. 2.13 in [3]), is employed

$$Q_n(s) = \frac{s}{c} + \frac{1}{2\rho} + \frac{s}{c} \frac{K'_n(\rho s/c)}{K_n(\rho s/c)} = \frac{1}{\rho} \left[ v + \frac{1}{2} + v \frac{K'_n(v)}{K_n(v)} \right], \quad v = \frac{s\rho}{c} \quad (1)$$

where  $K_n$  is the modified Bessel function of the second kind and  $n$ th order. The derivative  $K'_n$  of  $K_n$  is with respect to the argument  $\rho s/c$  where  $s$  is the Laplace domain variable,  $\rho$  is the radius of the fictitious circular nonreflecting boundary ( $\rho > 0$ ),  $c$  is the speed of light in the medium surrounding the cylinder ( $c > 0$ ), assumed here to be free space. From the scaling properties of the Laplace transform it is sufficient to expand the following expression in terms of partial functions

$$U_n(s) = s + \frac{1}{2} + s \frac{K'_n(s)}{K_n(s)} = s + \frac{1}{2} - s \frac{K_{n+1}(s)}{K_n(s)} + n \quad (2)$$

In (2), the derivative  $K'_n$  of  $K_n$  is now with respect to the argument  $s$ . Once the ILT of  $U_n(s)$  is found, denoted as  $u_n(t)$ , then the ILT of (1) can be obtained, for any  $\rho$  and  $c$ , as follows

$$q_n(t) = \frac{c}{\rho^2} u_n\left(\frac{c}{\rho} t\right) \quad (3)$$

The last expression in (2) is obtained using the identities of modified Bessel functions in [5] and it is the expression employed in VF with  $s = j\omega$ . In general, VF approximately expresses  $U_n(s)$  in (2) in the following form

$$U_n(s) \approx U_{n,app}(s) = d_n + h_n s + \sum_{m=1}^M \frac{r_{m,n}}{s - a_{m,n}} \quad (4)$$

where the parameters  $d_n$ ,  $h_n$ ,  $r_{m,n}$  (pole coefficient) and  $a_{m,n}$  (pole location) are computed by VECTFIT (version 1). The subscript ‘*app*’ indicates an approximation. To enable VECTFIT to produce the desirable results, the asymptotic value of  $U_n(s)$  as  $s \rightarrow 0$  is required. From [5], the small argument approximation of  $U_n(s)$  is

$$\lim_{s \rightarrow 0} U_n(s) \approx \frac{1}{2} - |n| \quad \forall n \in \mathbf{Z} \quad \text{since} \quad \lim_{s \rightarrow 0} K_n(s) \approx \frac{1}{2} \Gamma(n) \left( \frac{1}{2} s \right)^{-n} \quad \text{for } n > 0 \quad (5)$$

where  $\Gamma(\dots)$  is the Gamma function [5]. Furthermore, computing  $d_n$  and  $h_n$  is optional in VECTFIT and therefore knowledge of the large argument approximation of  $U_n(s)$  is beneficial. In this case,  $U_n(s) \rightarrow 0$  as  $s \rightarrow \infty$ , hence,  $d_n = h_n = 0$ . Therefore, after applying ILT to the partial fraction expansion in (4) one obtains the following approximate time domain expression for  $u_n(t)$ ,

$$u_{n,app}(t) = \sum_{m=1}^M r_{m,n} e^{a_{m,n} t} \quad (6)$$

A benefit from employing such a summation in a FETD-MNRBC formulation is that it will allow the convolution that appears in the boundary integral term of the formulation to be computed in a fast manner using a recursive approach as shown in [6] for 2D planar periodic structures.

*Numerical Results:* The input parameters used in VECTFIT were: (a) 4000 frequency samples; (b) a frequency range of  $0 \leq f \leq f_{max}$  with  $f_{max} = 4$  Hz; (c) iter = 20 iterations; and (d) asympflag = 1 (as  $d_n = h_n = 0$ ). Table 1 lists the computed partial fraction parameters,  $r_{m,n}$  and  $a_{m,n}$ . Their values are truncated to six decimal places. For  $n$  and  $M$  the values tabulated in [3] are used. Figure 1 shows plots of the function  $U_n(s = j2\pi f)$  versus frequency  $f$  based on: (a) the exact equation (2);

(b) the partial fraction approximation of the VF using the parameters in table 1 (without truncation); and (c) the partial fraction approximation of Alpert et al using the tabulated parameters in [3]. The absolute error value  $10\log_{10}|e_n(s=j2\pi f)|$  versus frequency  $f$  is also shown in figure 1 for both approximations, where

$$|e_n(s)| = \left| U_n(s) - \sum_{m=1}^M \frac{r_{m,n}}{s - a_{m,n}} \right| \quad (7)$$

The figure indicates that both approximations are of comparable accuracy and that  $10\log_{10}|e_n| < -50$  dB for  $n = 1, 2, 3, 4$ .

The VF accuracy enabled us to obtain, for the first time, FETD-MNRBC results (figure 2) based on the proposed VF approach. The plot in figure 2 shows the bistatic scattering width ( $\sigma_{2-D}/\lambda$ ), as defined in [7], of a perfectly electrically conducting triangular cylinder. The cylinder is surrounded by free space. The FETD-MNRBC results are compared with the integral equation formulation results presented in [2].

*Conclusion:* A VF approximation of a cylinder nonreflecting boundary kernel is proposed and validated. FETD-MNRBC results, based on this VF approximation, are presented for the first time.

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## Figure and Table Captions

### Figure 1

#### Caption:

Plots of  $|U_n(s = j2\pi f)|$  and  $10\log_{10}|e_n(s = j2\pi f)|$  versus frequency  $f$  for  $n = 1, 2, 3, 4$ .

### Figure 2

#### Caption:

Bistatic scattering width,  $10\log_{10}(\sigma_{2-D}/\lambda)$ , of a triangular cylinder.  $x_1 = 1.0\lambda$ ,  $x_2 = -0.707\lambda$ ,  $y_1 = 0.707\lambda$  and  $y_2 = -y_1$ . For simplicity,  $\lambda = 1$  m. The incident plane wave, with the electric field vector in the  $z$ -direction, is propagating in the negative  $x$ -direction. Surface  $\Omega$  represents the finite element region and  $\Gamma$  represents the circular boundary on which the MNRBC is applied.

### Table 1

#### Caption:

VF computed parameters

Figure 1

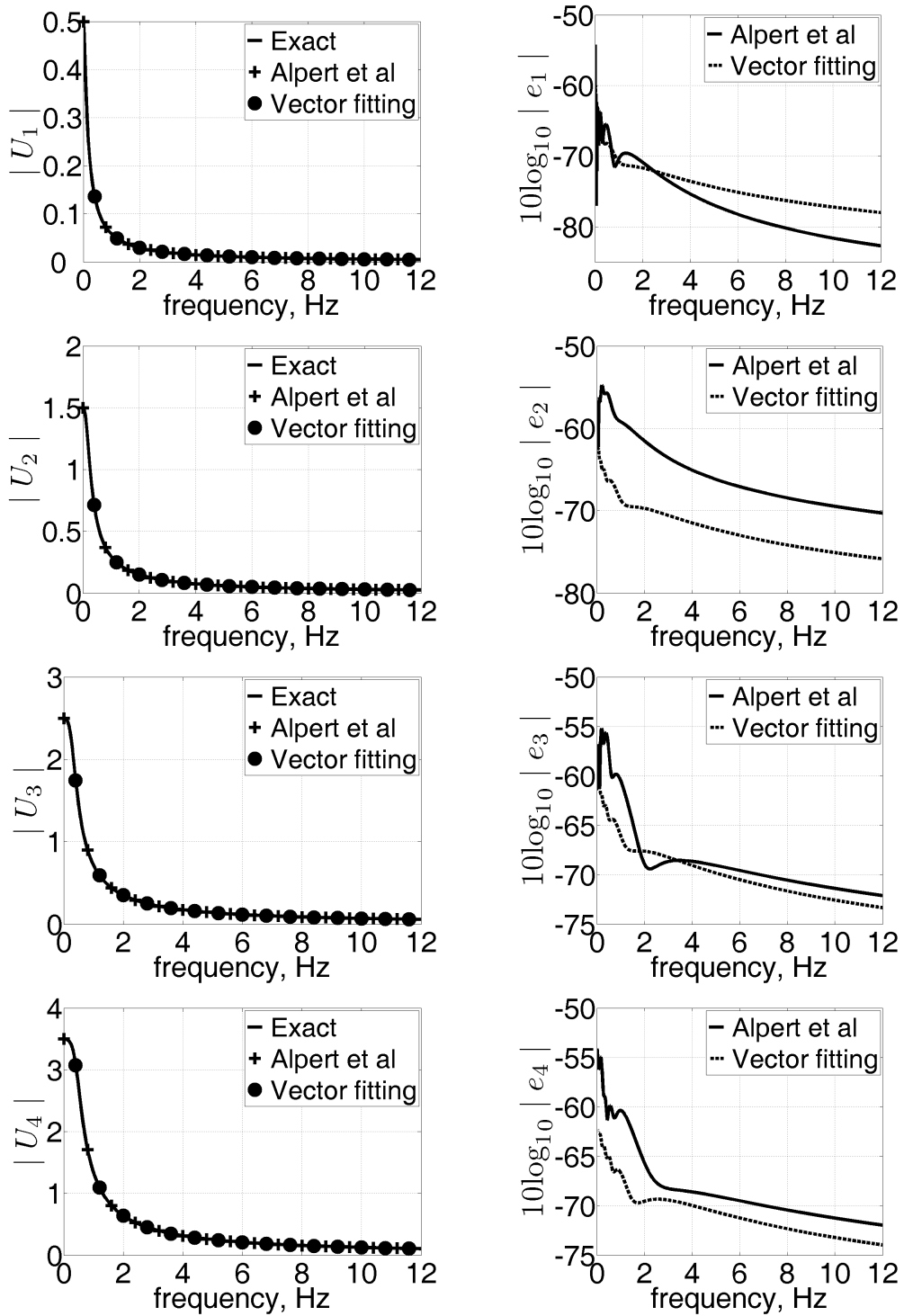
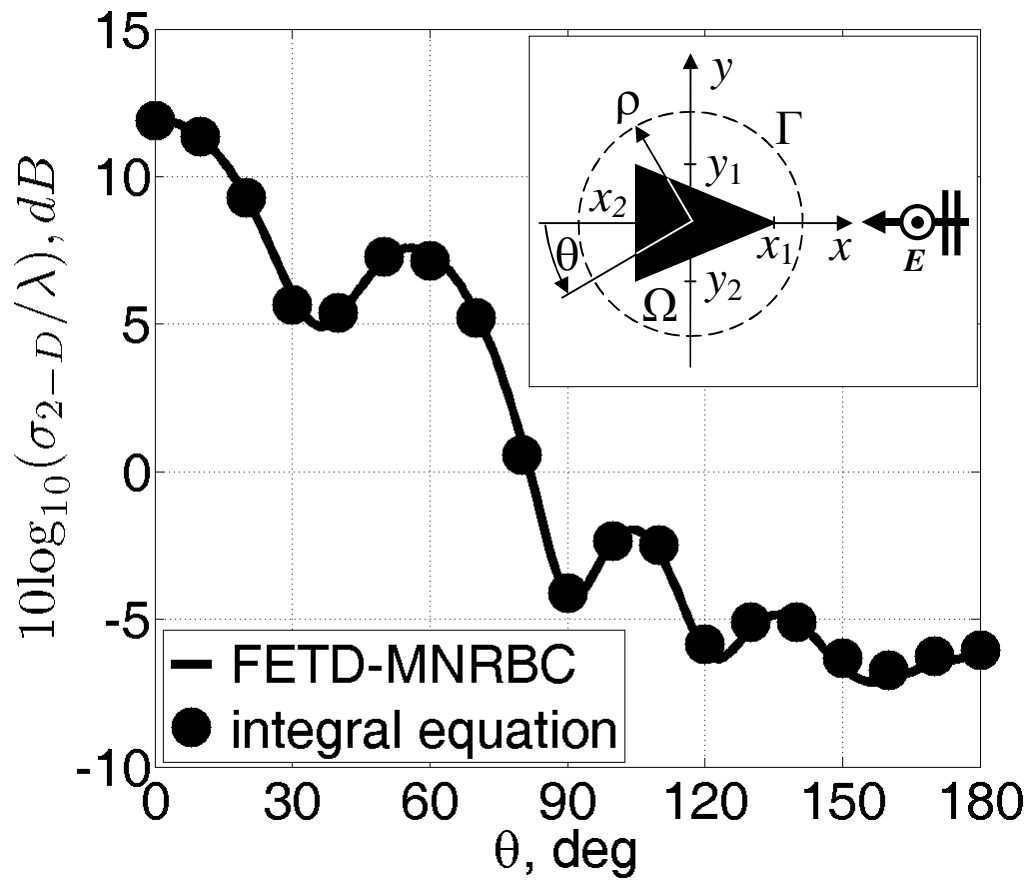




Figure 2



**Table 1**

$n$	$M$	Pole coefficients		Pole locations	
		Real	Imaginary	Real	Imaginary
1	9	$-6.125650 \times 10^{-3}$	0	$-3.465994 \times 10^0$	0
		$-5.259487 \times 10^{-2}$	0	$-1.904945 \times 10^0$	0
		$-1.381366 \times 10^{-1}$	0	$-1.091376 \times 10^0$	0
		$-1.326909 \times 10^{-1}$	0	$-6.548468 \times 10^{-1}$	0
		$-3.918820 \times 10^{-2}$	0	$-3.765733 \times 10^{-1}$	0
		$-5.643938 \times 10^{-3}$	0	$-1.921838 \times 10^{-1}$	0
		$-5.779033 \times 10^{-4}$	0	$-8.671415 \times 10^{-2}$	0
		$-3.945792 \times 10^{-5}$	0	$-3.308866 \times 10^{-2}$	0
		$-1.182999 \times 10^{-6}$	0	$-8.940158 \times 10^{-3}$	0
2	6	$2.087022 \times 10^{-4}$	0	$-2.352687 \times 10^{-1}$	0
		$1.883591 \times 10^{-2}$	0	$-6.004027 \times 10^{-1}$	0
		$9.779644 \times 10^{-1}$	0	$-1.585323 \times 10^0$	0
		$2.405753 \times 10^{-2}$	0	$-3.281031 \times 10^0$	0
		$-1.448034 \times 10^0$	$1.672191 \times 10^{-1}$	$-1.261094 \times 10^0$	$4.080800 \times 10^{-1}$
		$-1.448034 \times 10^0$	$-1.672191 \times 10^{-1}$	$-1.261094 \times 10^0$	$-4.080800 \times 10^{-1}$
3	5	$-1.096578 \times 10^{-2}$	0	$-9.316772 \times 10^{-1}$	0
		$-7.920391 \times 10^{-1}$	0	$-1.852993 \times 10^0$	0
		$-1.997077 \times 10^{-1}$	0	$-3.049055 \times 10^0$	0
		$-1.686141 \times 10^0$	$1.291524 \times 10^0$	$-1.680029 \times 10^0$	$1.307535 \times 10^0$
		$-1.686141 \times 10^0$	$-1.291524 \times 10^0$	$-1.680029 \times 10^0$	$-1.307535 \times 10^0$
4	5	$3.742610 \times 10^{-1}$	0	$-1.975139 \times 10^0$	0
		$-2.148009 \times 10^0$	$1.917512 \times 10^0$	$-2.813927 \times 10^0$	$4.063061 \times 10^{-1}$
		$-2.148009 \times 10^0$	$-1.917512 \times 10^0$	$-2.813927 \times 10^0$	$-4.063061 \times 10^{-1}$
		$-1.976622 \times 10^0$	$2.208657 \times 10^0$	$-1.978586 \times 10^0$	$2.204506 \times 10^0$
		$-1.976622 \times 10^0$	$-2.208657 \times 10^0$	$-1.978586 \times 10^0$	$-2.204506 \times 10^0$